

Design of Third-Order Tensorial RLS Adaptive Filtering Algorithms

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Introduction

- Conventional RLS algorithm \rightarrow high complexity

 Challenge → identification of long length impulse responses (e.g., network/acoustic echo paths)
large matrix → complexity / numerical issues

- In this paper: RLS adaptive filter using nearest Kronecker product (NKP) and third-order tensor decomposition
 (TOT) of the impulse response

Nearest Kronecker Product (NKP) Decomposition

! Solution for long-length filters:

 \rightarrow decomposition of impulse responses ($\otimes \rightarrow$ Kronecker product)

- Impulse response **h** of length $L = L_1 L_2$
- Low-rank systems (e.g., echo paths):

$$\mathbf{\hat{h}} = \sum_{p=1}^{P} \mathbf{h}_{2,p}^{L} \otimes \mathbf{h}_{1,p}^{L}, \quad P < L_2 \quad \Rightarrow \quad \mathbf{\hat{h}} = \sum_{p=1}^{P} \mathbf{\hat{h}}_{2,p}^{L} \otimes \mathbf{\hat{h}}_{1,p}^{L} \\ \text{decomposition parameter} \quad (estimate)$$

Decomposition-based approach: L = L₁L₂ → PL₁+ PL₂
→ reformulating a high-dimension system identification problem as a combination of low-dimension solutions.

[Paleologu, Benesty, Ciochină, "Linear system identification based on a Kronecker product decomposition," *IEEE Trans. Audio, Speech, Language Process.*, 2018]

NKP Decomposition (cont.)

• Impulse response **h** of length $L = L_1 L_2$



• Reshape vector $\mathbf{h} \rightarrow \mathbf{H}$ - matrix $L_1 \times L_2$



• NKP \iff singular value decomposition (SVD) of **H**

Third-Order Tensor (TOT) Decomposition

- Impulse response **h** of length $L = L_1 L_2 L_3$
- $\mathbf{h} = \sum_{p=1}^{P} \mathbf{h}_{3,p}^{L_3} \otimes \mathbf{h}_{2,p}^{L_2} \otimes \mathbf{h}_{1,p}^{L_1}, P = ?$ $\mathcal{H} = \sum_{p=1}^{P} \mathbf{h}_{1,p} \circ \mathbf{h}_{2,p} \circ \mathbf{h}_{3,p}$ $\mathbf{h} = \sum_{p=1}^{P} \mathbf{h}_{1,p} \circ \mathbf{h}_{2,p} \circ \mathbf{h}_{3,p}$ $\mathbf{h} = \sum_{p=1}^{P} \mathbf{h}_{1,p} \circ \mathbf{h}_{2,p} \circ \mathbf{h}_{3,p}$ $\mathbf{h} = \sum_{p=1}^{P} \mathbf{h}_{1,p} \circ \mathbf{h}_{2,p} \circ \mathbf{h}_{3,p}$
- ! Finding the rank of a TOT is a challenging task
- ! Challenge: avoiding approximation techniques

[Benesty, Paleologu, Ciochină, "Linear system identification based on a third-order tensor decomposition," *IEEE Signal Processing Letters*, 2023]

TOT Decomposition (cont.)

• Impulse response **h** of length $L = L_1 L_2$ with $L_1 >> L_2$ and $L_1 = L_{11} L_{12}$ $\mathbf{h} = \sum_{i=1}^{L_2} \mathbf{h}_2^i \otimes \mathbf{h}_1^i \qquad \mathbf{h}_1^i = \sum_{j=1}^{L_{12}} \mathbf{h}_{12}^{ij} \otimes \mathbf{h}_{11}^{ij} \qquad \mathbf{h} = \sum_{i=1}^{L_2} \sum_{j=1}^{L_{12}} \mathbf{h}_2^i \otimes \mathbf{h}_{12}^{ij} \otimes \mathbf{h}_{11}^{ij}$ • Consider that \mathbf{h}_1^i is low-rank $\longrightarrow \mathbf{h}_1^i = \sum_{j=1}^{P} \mathbf{h}_{12}^{ip} \otimes \mathbf{h}_{11}^{ip}$, $P < L_{12}$ $= \sum_{l=1}^{L_2} \sum_{p=1}^{P} \mathbf{h}_2^l \otimes \mathbf{h}_{12}^{lp} \otimes \mathbf{h}_{11}^{lp} \longleftrightarrow \mathcal{H} = \sum_{l=1}^{P} \sum_{p=1}^{L_2} \mathbf{h}_{11}^{lp} \circ \mathbf{h}_{12}^{lp} \circ \mathbf{h}_2^l$ $= \sum_{p=1}^{P} \mathcal{H}_p \quad (\text{sum of } P \text{ TOTs of rank } L_2) \quad \text{TOT of rank } L_2$ (no approximation)(no approximation) p = $\mathbf{h}(L) \Rightarrow \mathbf{h}_{2}^{l}(L_{2}^{2}) \& \mathbf{h}_{12}^{lp}(PL_{12}) \& \mathbf{h}_{11}^{lp}(PL_{11}) \qquad L_{2} \ll L_{11}L_{12}, P \ll L_{12}$ $L = L_{11}L_{12}L_2 \rightarrow L_2^2 + PL_{11} + PL_{12}$ (reduced number of parameters) 7

RLS Based on TOT

• Goal: "extract" / "separate" the individual components: $\mathbf{h}_2^l \& \mathbf{h}_{12}^{lp} \& \mathbf{h}_{11}^{lp} (l=1,...,L_2, p=1,...,P)$

• <u>Extraction</u> of \mathbf{h}_{2}^{l} $(l = 1, \dots, L_{2})$: $\mathbf{h} = \sum_{l=1}^{L_{2}} \sum_{p=1}^{P} \mathbf{h}_{2}^{l} \otimes \mathbf{h}_{12}^{lp} \otimes \mathbf{h}_{11}^{lp} = \sum_{l=1}^{L_{2}} \sum_{p=1}^{P} \left(\mathbf{I}_{L_{2}} \otimes \mathbf{h}_{12}^{lp} \otimes \mathbf{h}_{11}^{lp} \right) \mathbf{h}_{2}^{l}$ $= \sum_{l=1}^{L_{2}} \sum_{p=1}^{P} \mathbf{H}_{12,11}^{lp} \mathbf{h}_{2}^{l} = \sum_{l=1}^{L_{2}} \overline{\mathbf{H}}_{12,11}^{l} \mathbf{h}_{2}^{l} = \underline{\mathbf{H}}_{12,11} \mathbf{h}_{2}^{l}$

$$\mathbf{H}_{12,11}^{lp} = \mathbf{I}_{L_2} \otimes \mathbf{h}_{12}^{lp} \otimes \mathbf{h}_{11}^{lp}, \ \overline{\mathbf{H}}_{12,11}^{l} = \sum_{p=1}^{P} \mathbf{H}_{12,11}^{lp}$$

$$\overline{\mathbf{H}}_{12,11} = \begin{bmatrix} \overline{\mathbf{H}}_{12,11}^1 & \cdots & \overline{\mathbf{H}}_{12,11}^L \end{bmatrix}, \quad \underline{\mathbf{h}}_2 = \begin{bmatrix} \left(\mathbf{h}_2^1\right)^T & \cdots & \left(\mathbf{h}_2^{L_2}\right)^T \end{bmatrix}^T$$

RLS Based on TOT (cont.)

- <u>Extraction</u> of \mathbf{h}_{12}^{lp} & \mathbf{h}_{11}^{lp} $(l = 1, \dots, L_2, p = 1, \dots, P)$ $\mathbf{h} = \sum_{l=1}^{L_2} \sum_{p=1}^{P} \mathbf{h}_2^l \otimes \mathbf{h}_{12}^{lp} \otimes \mathbf{h}_{11}^{lp} = \sum_{l=1}^{L_2} \sum_{p=1}^{P} \left(\mathbf{h}_2^l \otimes \mathbf{I}_{L_{12}} \otimes \mathbf{h}_{11}^{lp}\right) \mathbf{h}_{12}^{lp} = \dots = \underline{\mathbf{H}}_{2,11} \underline{\mathbf{h}}_{12}$ $= \sum_{l=1}^{L_2} \sum_{p=1}^{P} \left(\mathbf{h}_2^l \otimes \mathbf{h}_{12}^{lp} \otimes \mathbf{I}_{L_{11}}\right) \mathbf{h}_{11}^{lp} = \dots = \underline{\mathbf{H}}_{2,12} \underline{\mathbf{h}}_{11}$
- <u>Notation</u>: \mathbf{g}_* and $\mathbf{G}_* \rightarrow \text{estimates of } \mathbf{h}_*$ and \mathbf{H}_* , respectively $e(n) = d(n) - \mathbf{g}_2^T(n-1)\mathbf{x}(n)$ \Longrightarrow Least-squares (LS) criterion $= d(n) - \mathbf{g}_2^T(n-1)\mathbf{\overline{G}}_{12,11}^T(n-1)\mathbf{x}(n) = d(n) - \mathbf{g}_2^T(n-1)\mathbf{x}_{12,11}(n) = e_1(n)$ $= d(n) - \mathbf{\overline{g}}_{12}^T(n-1)\mathbf{\overline{G}}_{2,11}^T(n-1)\mathbf{x}(n) = d(n) - \mathbf{\overline{g}}_{12}^T(n-1)\mathbf{x}_{2,11}(n) = e_2(n)$ $= d(n) - \mathbf{\overline{g}}_{11}^T(n-1)\mathbf{\overline{G}}_{2,12}^T(n-1)\mathbf{x}(n) = d(n) - \mathbf{\overline{g}}_{11}^T(n-1)\mathbf{x}_{2,12}(n) = e_3(n)$

RLS Based on TOT (cont.)

- LS cost functions Normal equations $\begin{cases} J\left(\underline{\mathbf{g}}_{2} \middle| \overline{\underline{\mathbf{g}}}_{12}, \overline{\underline{\mathbf{g}}}_{11}\right) \implies \mathbf{R}_{12,11}(n) \underline{\mathbf{g}}_{2}(n) = \mathbf{r}_{12,11}(n) \implies \underline{\mathbf{g}}_{2}(n) \\ J\left(\overline{\underline{\mathbf{g}}}_{12} \middle| \underline{\mathbf{g}}_{2}, \overline{\underline{\mathbf{g}}}_{11}\right) \implies \mathbf{R}_{2,11}(n) \overline{\underline{\mathbf{g}}}_{12}(n) = \mathbf{r}_{2,11}(n) \implies \overline{\underline{\mathbf{g}}}_{12}(n) \\ J\left(\overline{\underline{\mathbf{g}}}_{11} \middle| \underline{\mathbf{g}}_{2}, \overline{\underline{\mathbf{g}}}_{12}\right) \implies \mathbf{R}_{2,12}(n) \overline{\underline{\mathbf{g}}}_{11}(n) = \mathbf{r}_{2,12}(n) \implies \overline{\underline{\mathbf{g}}}_{11}(n) \end{cases}$ fixed \rightarrow RLS adaptive algorithm using TOT decomposition (RLS-TOT)
- Final estimate:

l=1 *p*=1

Advantages

- smaller data structures (matrices)
- faster convergence/tracking
- $\mathbf{g}(n) = \sum_{l=1}^{L_2} \sum_{l=1}^{P} \mathbf{g}_2^l(n) \otimes \mathbf{g}_{12}^{lp}(n) \otimes \mathbf{g}_{11}^{lp}(n)$ - lower computational complexity

Simulation Results

- Conditions:
 - → h from ITU-T Rec. G168, with L = 512.

→ TOT decomposition: $L_{11} = L_{12} = 16$, $L_2 = 2$

 \rightarrow h acoustic impulse response, with *L* = 2048.

→ TOT decomposition: $L_{11} = L_{12} = 32$, $L_2 = 2$

- \rightarrow input signal AR(1) process with pole at 0.8 / speech sequence
- \rightarrow additive noise white Gaussian noise, with SNR = 20 or 10 dB.
- \rightarrow performance measure: normalized misalignment (dB).

$$20\log_{10}\left[\left\|\mathbf{h}-\mathbf{g}(n)\right\|_{2}/\left\|\mathbf{h}\right\|_{2}\right]$$

- Algorithms:
- \rightarrow conventional **RLS**
- \rightarrow RLS-TOT
- \rightarrow RLS-NKP / APA / DR-FRLS (see the references [6] / [12] / [13])

Simulation Results (cont.)



Figure 1. Complexity order of the conventional RLS algorithm and RLS-TOT for two impulse responses, with lengths (a) L = 512 and (b) L = 2048.

Simulation Results (cont.)



Figure 2. Misalignment of the RLS-based algorithms for the identification of a network impulse response of length L = 512. The forgetting factors are set based on equation (4), using K = 5 for the conventional RLS algorithm, and K = 45 for the RLS-NKP and RLS-TOT. The input signal is an AR(1) process and SNR = 20 dB.

Simulation Results (cont.)



Figure 3. Misalignment of the APA, DR-FRLS algorithm, and RLS-TOT, for the identification of an acoustic impulse response of length L = 2048. The RLS-TOT uses two forgetting factors set based on equation (4), with K = 100, while the third one is equal to 1. The input signal is speech and SNR = 10 dB.

Simulation Results (cont.)



Figure 4. Impulse responses related to the experiment reported in Figure 3: (a) true acoustic impulse response **h**; (b) the estimate obtained by APA using the step-size equal to 1; (c) the estimate obtained by DR-FRLS using the data-reuse parameter equal to 12; and (d) the estimate obtained by RLS-TOT using P = 8.

Conclusions and Perspectives

- System identification exploiting a third-order tensor (TOT) decomposition.
- Efficient solution for the identification of long-length low-rank systems (e.g., echo paths).
- High-dimension system identification problem → reformulated as a combination of low-dimension solutions (three shorter filters).
- *Solution*: RLS adaptive filter based on TOT \rightarrow RLS-TOT.
- The RLS-TOT outperforms the conventional RLS and other RLS-based algorithms (faster convergence/tracking & lower computational complexity).
- **Future works**: dichotomous CD (DCD) \rightarrow reduce complexity.
- extension to multidimensional case \rightarrow higher-order tensors.
- improved versions with variable forgetting factors and variable regularization parameters.

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